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Elect/Elect

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## Question

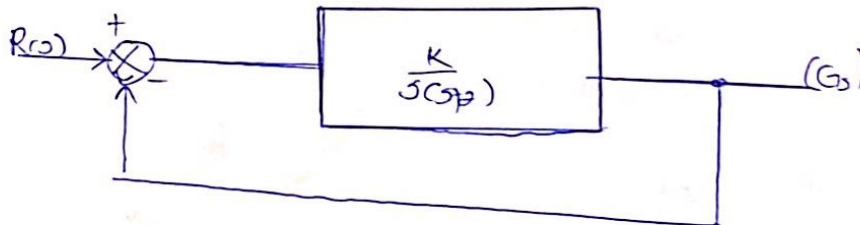
- ① Briefly explain root locus technique
- ② Describe the use Routh's H rule to find the stability of a closed loop system when: Entire row is zero on the routh table  
To determine the poles on the  $j\omega$  axis

## Solution

Root locus technique :- Root locus is a closed loop poles path by varying the gain  $K$  from zero to infinity

- locus path
- Root means root of characteristic equation  $C(s)$  of the closed loop

• Root locus is a graphical method in which roots of the characteristic equation are plotted in the  $s$ -plane for the different values of parameter, the locus of the roots of the characteristic equation when the gain is varied from zero to infinity. Consider a unity feedback as shown below:



$$\frac{C(s)}{R(s)} \text{ where } G(s) = \frac{K}{s(s+2)} ; H(s) = 1$$

$$1 + \frac{K}{s(s+2)} - 1 = 0$$

$\therefore$  The root of the equation

$$s_1 = -1 + \sqrt{1-K} , s_2 = -1 - \sqrt{1-K}$$

As ' $K$ ' is varied, the two roots given the loci in  $s$ -plane for the various values of  $K$ , the location of the roots are

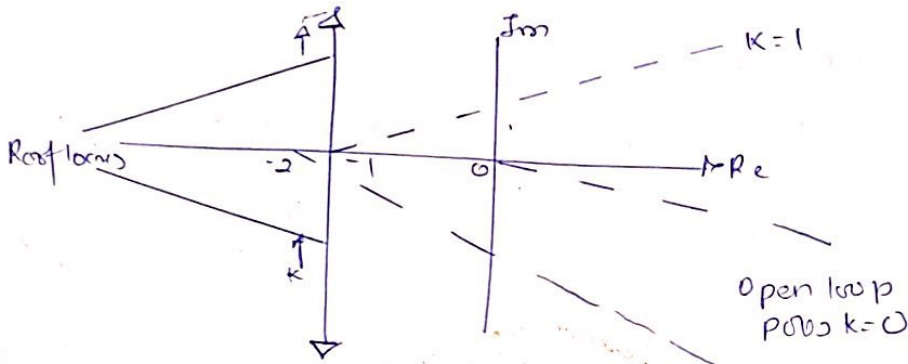
- ① When  $0 < K < 1$ , the roots are real and distinct
- ② When  $K = 0$ , the two roots are  $s_1 = 0$  and  $s_2 = 0$  (open loop poles)
- ③ When  $K = 1$  both roots are real and equal
- ④ When  $K > 1$  the roots are complex conjugate with real part = 1

\* Plotting root of a characteristic equation

\*  $k=0$ ;  $s=0$  &  $s=-2$

\*  $k=1$ ; Both root  $s=-1$

\*  $k > 1$ ; the roots break away from the real axis and become complex conjugate having negative real part equal to  $-1$ ;  $s_{1,2} = -1 \pm j\sqrt{k-1}$



③ Describe the use of Routh Hurwitz to find the stability of a closed loop system  
 Routh Hurwitz method is a quick mathematical way to determine what gain or allowed values of achieving stability. It doesn't show the path of the poles but it tells the stability of a system and its relation to a gain factor. This method can tell how many closed loop system poles are in the left half plane in the right half plane and on the  $j\omega$  axis.

\* Two methods required in Routh Hurwitz

① Operate a data table called Routh table

② Interpret the routh table to tell how many closed loop systems poles are in the left half plane the right half plane and on the  $j\omega$  axis

NB If the characteristic equation of the table is:  $a_5s^5 + a_4s^4 + a_3s^3 + a_2s^2 + a_1s + a_0 = 0$

$$\begin{matrix} s^5 & a_5 & a_3 & a_1 \\ s^4 & a_4 & a_2 & a_0 \\ s^3 & b_1 & b_2 & b_3 \\ s^2 & c_1 & c_2 & c_3 \end{matrix}$$

where  $b_1 = -\frac{a_1 a_4 - a_0 a_5}{a_4}$

$b_2 = \frac{a_1 a_3 - a_0 a_4}{a_4}$

$b_3 = \frac{a_1 a_2 - a_0 a_3}{a_4}$

$c_1 = \frac{a_2 a_4 - b_1 b_3}{b_1}$

$c_2 = \frac{a_2 a_3 - b_1 b_2}{b_1}$

$c_3 = \frac{a_2 a_0 - b_1 b_3}{b_1}$

Special case: Entire Row in zero on the Routh table

In this case, the entire row consists of zero because there is an even polynomial that is a factor of the original  $\rightarrow$  indicates the equation has at least one pair of roots which lie vertically opposite each other and equidistant from origin. The array can be completed by forming the auxiliary polynomial.

The polynomial whose coefficients are elements of the row just above the zeros in Routh array is called an auxiliary polynomial.

An example will be used to demonstrate how to construct and interpret the Routh table.

Consider the following characteristic equation

$$s^6 + 25s^5 + 155s^4 + 105s^3 + 205s^2 + 165s + 16 = 0$$

$s_6$	1	8	20	16
$s_5$	2	12	16	0
$s_4$	2	12	10	0
$s_3$	0	0	0	0
$s_2$	0	0	0	0
$s_1$	0	0	0	0

From the above table it is clear that the fourth row is zero. Hence the auxiliary equation,  $f(s)$  can be formed from the coefficients of the row just above the particular row.

Auxiliary equation:  $f(s) = 25s^4 + 105s^3 + 16 = 0$   
 $\therefore \frac{d f(s)}{ds} = 105s^3 + 24s$

Once the order of characteristic eqn is six in order of eqn for the two rows  $(s_4)$  &  $(s_5)$  to be left half of the s-plane and four roots of auxiliary eqn are dominant.

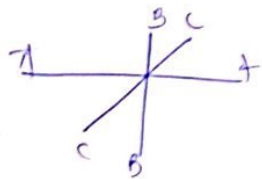
$s_6$	1	8	20	16
$s_5$	2	12	16	0
$s_4$	2	12	16	0
$s_3$	8	24	0	0
$s_2$	16	16	0	0
$s_1$	20	0	0	0

Result

- ① No sign change in first column
- ② The system is stable (marginal stability)
- ③ Poles lie on the imaginary axis
- ④ The two roots lie on the left half the s-plane
- ⑤ The four roots are dominant

\* To determine the poles on the  $s$  axis

The poles on the  $s$  axis means that there is a row of zeros and closed loop system with only imaginary axis poles & nothing and poles in the left half plane if is marginally stable. The imp of such system is imaginary and symmetrical about the origin



From the example above: Determine the poles of the  $G(s)$

Since it is the row of zeros it means the roots are lying on the imaginary axis the poles can be determined by equating the denominator to zero;  $f(s) = 0$

$$\therefore f(s) : s^4 + 10s^2 + 16 = 0$$

$$\text{let } s^2 = x$$

$$\therefore 2x^2 + 10x + 16 = 0$$

$$2x^2 + 10x + 16 = 0$$

$$x^2 + 5x + 8 = 0$$

$$(x+2)(x+4) = 0$$

$$x+2 = 0 \quad x+4 = 0$$

$$x = -2 \quad x = -4$$

$$\text{Recall: } x = s^2 \quad s^2 = -4$$

$$\therefore s^2 = -2 \quad s^2 = -\sqrt{-1}$$

$$s = \pm \sqrt{-2} \quad s = \pm j\sqrt{2}$$

$$s = \pm j\sqrt{2} \quad s = \pm j\sqrt{2}$$

$\therefore$  The poles on the imaginary axis are given as  $\pm j\sqrt{2} \pm 2$

Since the roots are non-repeated on the imaginary axis, hence the system marginally stable

No 4 roots with positive real part = 0

No 4 roots with negative real part = 2

No 4 roots with zero real part = 4